

Inductively Coupled Thermistor

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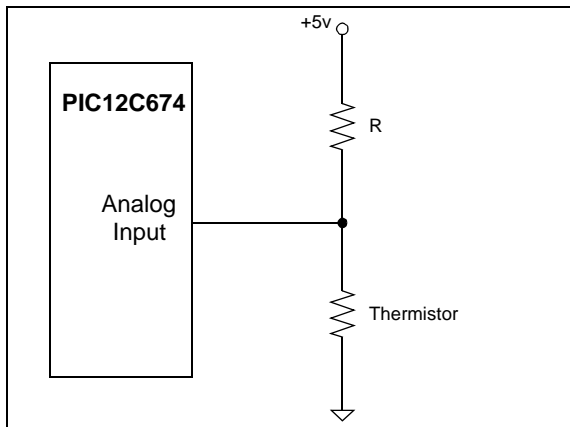
INTRODUCTION

This application note describes a temperature sensor using a PICmicro[®] microcontroller. This temperature sensor is unique because it can work through non-ferromagnetic walls. Interfacing to a thermistor is normally a very straightforward task. Interfacing to a thermistor on the other side of a wall gets a little more complicated. Interfacing through walls can be a very valuable feature in certain applications, notably temperature sensing for refrigerators or sensing inside of a hazardous gas environment where an isolation barrier is a safety issue.

Implementation

Thermistors come in all types and values. For temperature sensing, most applications call for a NTC (negative temperature coefficient) type of thermistor where the resistance goes down with increasing temperature. Naturally, the resistance does not go down linearly, so some processing is required to translate the resistance to a temperature. This is of course a perfect job for a PICmicro microcontroller. The normal method of interfacing to a thermistor is shown below.

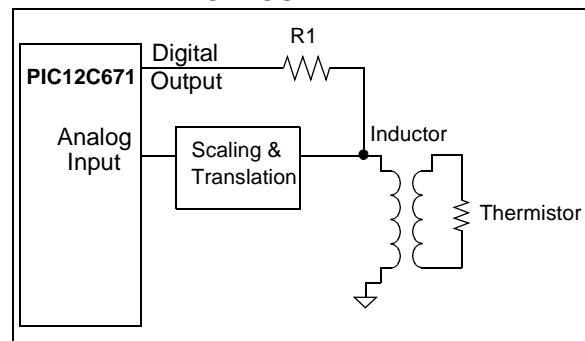
FIGURE 1: BASIC THERMISTOR CIRCUIT



This method is inappropriate for a through wall temperature system because it requires a wire to connect the thermistor to the PICmicro MCU. DC currents cannot be transmitted through a wall, so this method cannot be directly converted to a through wall system.

To sense through the wall we must get current flowing through the wall. This is easily done with a pair of coils of sufficient diameter to couple through the wall. About 100 winds of wire around a 12cm disk (size of CD) will provide sufficient inductance and size to couple through ½ inch. The PICmicro MCU can source up to 25 mA so that is how the coil will be energized. See the schematic below for the new circuit.

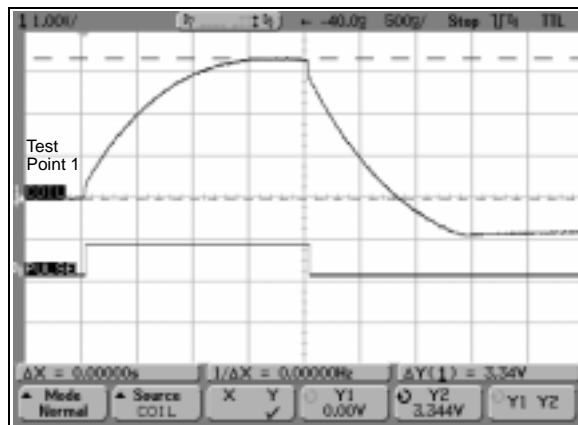
FIGURE 2: BLOCK DIAGRAM OF INDUCTIVELY COUPLED SENSOR



With each side of the transformer on different sides of the wall, we have created an isolation transformer. A thermistor is loading the secondary while the PICmicro MCU is driving the primary. Now the PICmicro MCU just needs a way to detect the voltage changes of the primary due to the loading on the secondary. One method is to feed a long pulse train to the coil and look at the amplitude modulation caused by the thermistor. This is effective, but it has two problems. Problem number one is, the continuous current through the thermistor will heat it and cause errors in the measurement. The second problem is, the circuitry to get a clean reading is complex and therefore not cheap.

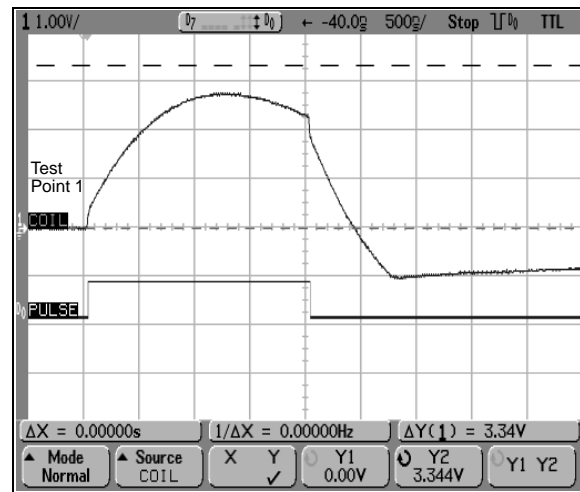
A preferred method is to send a single pulse into the inductor and look at the transient response of the coil. We can then simplify the circuitry and eliminate the self-heating. By placing a capacitor and resistor across the primary, a RLC type circuit is formed (see Appendix B). This circuit will have a gentle peak when we energize it. See Figure 3.

FIGURE 3: RLC PULSE RESPONSE



The idea is to have the initial capacitor current dominate the first microsecond so the circuit does not look like the infinite resistance to the coil. As the capacitor charges, the coil current starts to become the dominant factor and the load becomes visible as a variation in coil peak voltage. The RC values must be adjusted to maintain a ratio of approximately 10000:1 with the inductor. With this configuration, a load on the secondary coil causes a change in the peak voltage on the primary. See Figure 4, set for the same scale as Figure 3.

FIGURE 4: INDUCTIVELY LOADED RLC PULSE RESPONSE



The peak voltage changes with the load voltage. This is a very useful feature. To make a good measurement without an amazingly fast analog-to-digital converter we can add a sample and hold circuit. Because the voltage variation is very small (about 500 mV in this example) we need a gain of about 10 to get a 5V range. The 2.75V offset needs to be removed so a difference amplifier is used to subtract the offset and then multiply the gain. Appendix B shows the test circuit schematic.

The calibration for this circuit is simple. Adjust the pot with no load on the secondary inductor. Adjust the pot until the analog-to-digital converter (ADC) is no longer reading a full-scale voltage. Any load on the secondary will cause the voltage to drop. The gain should be set to get a good reading over the desired range.

Theory of Operation

The PICmicro MCU sends a pulse to the inductor, which induces a voltage in the secondary coil. The secondary voltage across the thermistor causes a current, which is seen as a voltage drop, on the primary. The larger the secondary current, the larger the voltage drop at the primary. The first Op Amp, U1A, implements a high speed peak hold circuit by only passing current that charges the capacitor, but not allowing the capacitor to discharge. The second Op Amp, U1B, buffers the capacitor to the difference amplifier. This prevents the capacitor voltage from dropping too fast. The third Op Amp, U1C, subtracts the offset voltage and multiplies the difference by a gain of 10. The offset voltage is provided by the fourth Op Amp, U1D. The result is read by the ADC. The capacitor (C2) is drained between reads by an output from the PICmicro MCU.

The diode prevents the PICmicro MCU from charging the capacitor. By a small change in the software, this diode could be eliminated if the PICmicro MCU pin were left as an input pin at high impedance until the capacitor needed discharging. Alternatively, the pin would not be required at all if a suitable load resistor were provided for the capacitor. This resistor would have to be large enough that the capacitor did not drain too much before the ADC sample period passed and small enough to drain the capacitor between measurements. Here is the code for a PIC12C67X that takes a measurement.

```
measure          ; do the measurement
  bcf  INTCON,GIE  ; disable irq's
  bsf  GPIO,holdcap ; arm the cap
  bsf  GPIO,coil   ; charge the coil
  nop                    ; wait a bit
  bcf  GPIO,coil   ; Turn off coil
  bsf  ADCON0,GO   ; start ADC
  btfsc ADCON0,GO  ; wait for ADC
  goto $-1         ;
  bcf  GPIO,holdcap ; dump the cap
  bsf  INTCON,GIE  ; enable irq's
  movf ADRES,W     ; result to W
  return          ; all done
```

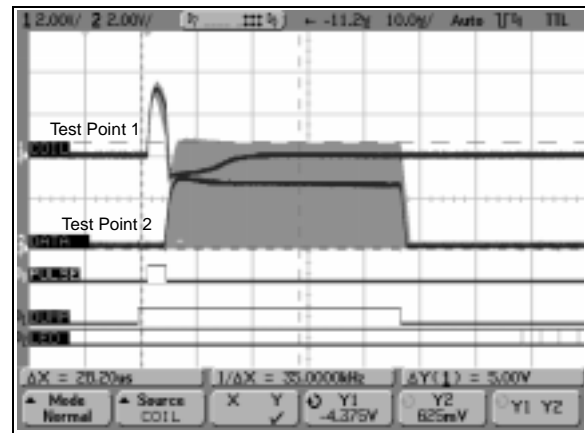
The slowest part of the measurement is waiting for the ADC to finish. In the test system, GPIO4 was used to drive an LED with a PWM signal. This PWM was generated with a Timer0 interrupt. To prevent the Timer0 interrupt from affecting the pulse timing, all interrupts are disabled during the critical section of the measurement code. The PICmicro MCU is operating from its internal RC oscillator. This leaves a few pins to accomplish other tasks.

Figure 5 illustrates the complete circuit performance using a 10k pot in place of a thermistor. The offset was adjusted until the input to the ADC was 5V without the secondary coil in place and without clipping. With the coil in place, the resistor was swept over its entire range and produces values inside the gray area. With a suitable scaling table, this output could easily be converted to a resistance or a temperature.

Conclusion

Using inductive coupling is common with keyless entry, low frequency RF and power supplies. This application note shows that inductive pulse coupling can also be effectively used to transfer information, like temperature sensing, through a non-ferromagnetic barrier.

FIGURE 5: RANGE OF RESPONSE AT THE OUTPUT



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APPENDIX A: CODE LISTING

```

;*****
;
;   Filename:      fridge.asm
;   Date:         10/13/2000
;   File Version:  1.0
;
;   Author:       Joseph Julicher
;   Company:      Microchip Technology
;
;*****

list      p=12ce674           ; list directive to define processor
#include <p12ce674.inc>       ; processor specific variable definitions
errorlevel -302              ; suppress message 302 from list file

__CONFIG  _CP_OFF & _WDT_OFF & _MCLRE_OFF & _PWRTE_ON & _INTRC_OSC_NOCLKOUT

;***** VARIABLE DEFINITIONS
w_temp    EQU    0x70         ; variable used for context saving
status_temp EQU    0x71      ; variable used for context saving

tsr        EQU    0x72         ; transmit shift register
bitcount   EQU    0x73         ; transmit bit counter
led         EQU    0x74         ; LED brightness
counter     EQU    0x75         ; LED PWM counter
temp        EQU    0x76         ; holding for PWM status

;***** CONSTANTS DEFINITIONS
speed      EQU    0xDf         ; PWM period constant

;***** PIN DEFINITIONS
holdcap    EQU    0x02         ; GPIO pin for the hold cap
coil       EQU    0x01         ; GPIO pin for the coil
pwm        EQU    0x04         ; GPIO pin for the pwm (LED brightness)

;*****
ORG        0x000              ; processor reset vector
goto      main                ; go to beginning of program

ORG        0x004              ; interrupt vector location
movwf     w_temp              ; save off current W register contents
movf      STATUS,w            ; move status register into W register
movwf     status_temp         ; save off contents of STATUS register

incf      counter,w          ; PWM routine
addwf     led,w
btfss    STATUS,C
bcf      GPIO,pwm
btfsc    STATUS,C

```

```

    bsf    GPIO,pwm
    movwf  counter

    bcf    INTCON,T0IF    ; clear the TMR0 flag
    bsf    INTCON,T0IE    ; reenable TMR0 interrupt

    movlw  speed
    movwf  TMR0

    movf   status_temp,w  ; retrieve copy of STATUS register
    movwf  STATUS         ; restore pre-isr STATUS register contents
    movf   w_temp,w       ; restore W register
    retfie                ; return from interrupt

main
    call   0x7FF          ; retrieve factory calibration value
    bsf    STATUS,RP0     ; set file register bank to 1
    movwf  OSCCAL         ; update register with factory cal value
    bcf    STATUS,RP0     ; set file register bank to 0

    clrf   TMR0          ; clear the timer
    clrf   counter
    clrf   led

; setup GPIO
    clrf   GPIO          ; set all I/O's to 0
    clrf   INTCON        ; clear all flags and enables
    bsf    INTCON,T0IE    ; enable TMR0 interrupt
    bsf    INTCON,GIE     ; enable all interrupts

    bsf    STATUS, RP0    ; Select Page 1
    clrf   OPTION_REG     ; clear all options
    bsf    OPTION_REG,NOT_GPPU; Turn off weak pullup

    movlw  B'00001001     ; GPIO 0 is Input
                                ; GPIO 1 is Output
                                ; GPIO 2 is Output
                                ; GPIO 3 is Input
                                ; GPIO 4 is Output
                                ; GPIO 5 is Output
    movwf  TRISIO

    movlw  B'00000110     ; GP0 is analog, VREF is Vdd
    movwf  ADCON1         ; Configure A/D Inputs

    bcf    PIE1,ADIE     ; disable A/D Interrupts

    bcf    STATUS, RP0    ; Select Page 0
    movlw  B'01000001     ; 8 Tosc clock, A/D is on, Channel 0 is selected
    movwf  ADCON0         ;
    bcf    PIR1, ADIF     ; Clear A/D interrupt flag bit

repeat    call   measure   ; make a measurement
          movwf  led       ; set the LED brightness
          movlw  D'56      ; wait 200 loops or 1ms
delay     nop             ;
          addlw  D'1        ;
          btfss STATUS,Z  ;
          goto  delay      ;
          goto  repeat     ;

measure   ; do the measurement
          bcf    INTCON,GIE ; disable all interrupts
          bsf    GPIO,holdcap ; arm the cap
          bsf    GPIO,coil   ; charge the coil
          nop
          bcf    GPIO,coil   ; Turn off coil

```

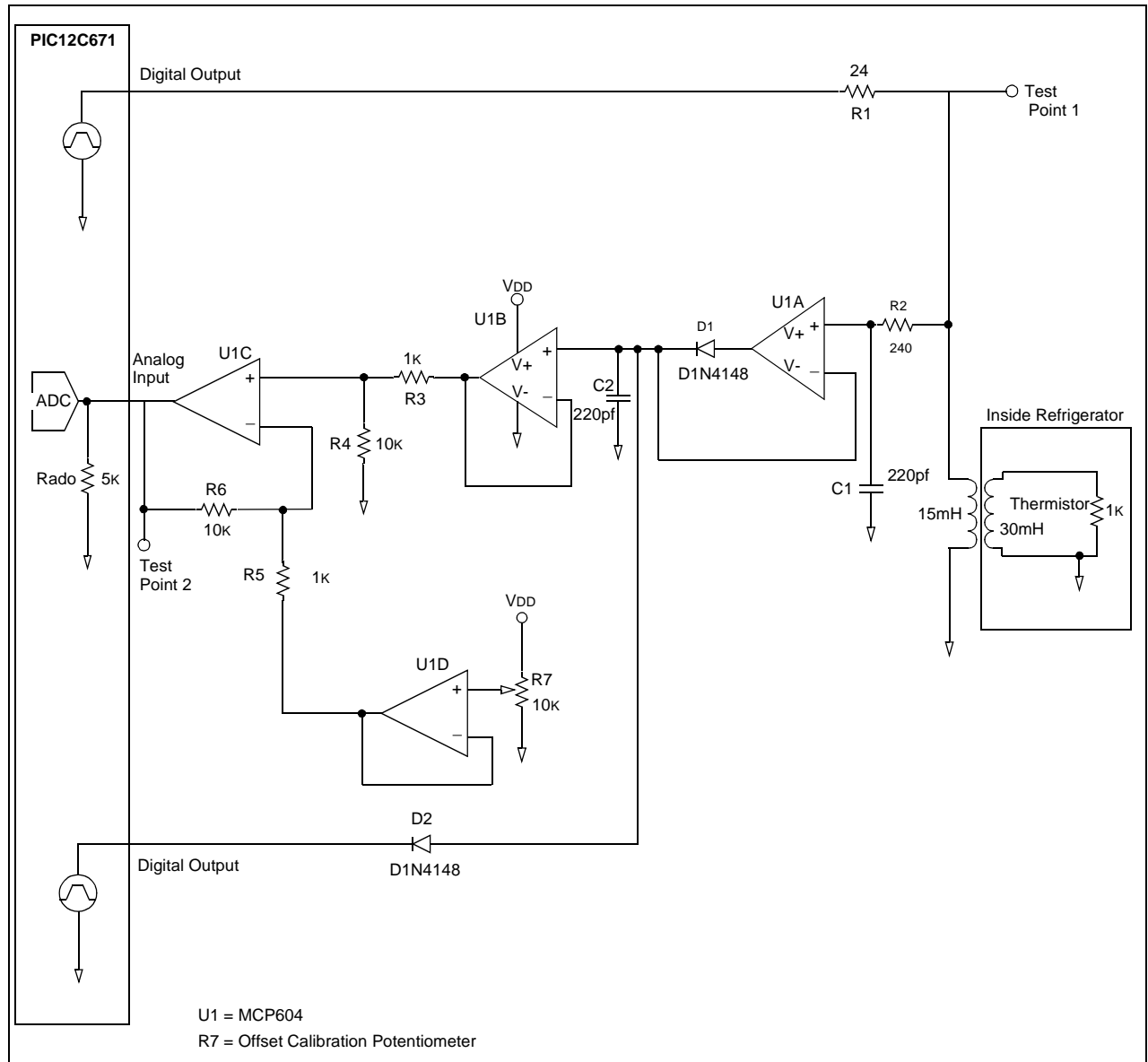
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```
                                ; wait for the inductor collapse to finish
bsf    ADCON0,GO                ; start ADC
btfsc  ADCON0,GO                ; wait for ADC to finish
goto   $-1                      ; go back if not finished yet
bcf    GPIO,holdcap             ; dump the cap
bsf    INTCON,GIE               ; enable all interrupts
movf   ADRES,W                  ; move the result to W
return                                ; all done
```

END

APPENDIX B: TEST SCHEMATIC

TEST SCHEMATIC



AN850

NOTES:

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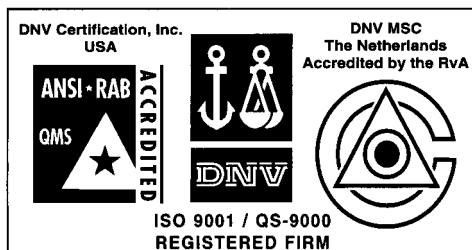
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